## Appendix M: Root Locus Rules: Derivations



## .1 Derivation of the Behavior of the Root Locus at Infinity (Kuo, 1987)

Let the open-loop transfer function be represented as follows:

$$KG(s)H(s) = \frac{K(s^m + a_1s^{m-1} + \dots + a_m)}{(s^{m+n} + b_1s^{m+n-1} + \dots + b_{m+n})}$$
(M.1)

or

$$KG(s)H(s) = \frac{K}{(\frac{s^{m+n} + b_1 s^{m+n-1} + \dots + b_{m+n}}{s^m + a_1 s^{m-1} + \dots + a_m})}$$
(M.2)

Performing the indicated division in the denominator, we obtain

$$KG(s)H(s) = \frac{K}{s^n + (b_1 - a_1)s^{n-1} + \cdots}$$
 (M.3)

In order for a pole of the closed-loop transfer function to exist,

$$KG(s)H(s) = -1 \tag{M.4}$$

Assuming large values of s that would exist as the locus moves toward infinity, Eq. (M.3) becomes

$$s^{n} + (b_{1} - a_{1})s^{n-1} = -K$$
(M.5)

Factoring out  $s^n$ , Eq. (M.5) becomes

$$s^n \left( 1 + \frac{b_1 - a_1}{s} \right) = -K \tag{M.6}$$

Taking the *n*th root of both sides, we have

$$s\left(1+\frac{b_1-a_1}{s}\right)^{1/n} = -K^{1/n}$$
 (M.7)

-		
-	1	
	1	
	1	

2

Appendix M: Root Locus Rules: Derivations

If the term

$$\left(1 + \frac{b_1 - a_1}{s}\right)^{1/n} \tag{M.8}$$

is expanded into an infinite series where only the first two terms are significant,<sup>1</sup> we obtain

$$s\left(1+\frac{b_1-a_1}{ns}\right) = (-K)^{1/n}$$
 (M.9)

Distributing the factor S on the left-hand side yields

$$s + \frac{b_1 - a_1}{n} = (-K)^{1/n} \tag{M.10}$$

Now, letting  $s = \sigma + j\omega$  and  $(-K)^{1/n} = |K^{1/n}|e^{j(2k+1)\pi/n}$ , where

$$(-1)^{1/n} = e^{j(2k+1)\pi/n} = \cos\left(\frac{(2k+1)\pi}{n}\right) + j\sin\left(\frac{(2k+1)\pi}{n}\right)$$
(M.11)

Eq. (M.10) becomes

$$\sigma + j\omega + \frac{b_1 - a_1}{n} = \left| K^{1/n} \right| \left[ \cos \frac{(2k+1)\pi}{n} + j \sin \frac{(2k+1)\pi}{n} \right]$$
(M.12)

where  $k = 0, \pm 1, \pm 2, \pm 3, \ldots$  Setting the real and imaginary parts of both sides equal to each other, we obtain

$$\sigma + \frac{b_1 - a_1}{n} = \left| K^{1/n} \right| \cos \frac{(2k+1)\pi}{n}$$
(M.13a)

$$\omega = |k^{1/n}|\sin - \frac{(2k+1)\pi}{n}$$
 (M.13b)

Dividing the two equations to eliminate  $|K^{1/n}|$ , we obtain

$$\frac{\sigma + \frac{b_1 - a_1}{n}}{\omega} = \frac{\cos\frac{(2k+1)\pi}{n}}{\sin\frac{(2k+1)\pi}{n}}$$
(M.14)

Finally, solving for  $\omega$ , we find

$$\omega = \left[\tan\frac{(2k+1)\pi}{n}\right] \left[\sigma + \frac{b_1 - a_1}{n}\right] \tag{M.15}$$

The form of this equation is that of a straight line,

$$\omega = M(\sigma - \sigma_0) \tag{M.16}$$

where the slope of the line, M, is

$$M = \tan\frac{(2k+1)\pi}{n} \tag{M.17}$$

<sup>&</sup>lt;sup>1</sup> This is a good approximation since s is approaching infinity for the region applicable to the derivation.

## M.2 Derivation of Transition Method for Breakaway and Break-in Points

Thus, the angle of the line in radians with respect to the positive extension of the real axis is

$$\theta = \frac{(2k+1)\pi}{n} \tag{M.18}$$

and the  $\sigma$  intercept is

$$\sigma_0 = -\left\lfloor \frac{b_1 - a_1}{n} \right\rfloor \tag{M.19}$$

From the theory of equations,<sup>2</sup>

$$b_1 = -\sum$$
 finite poles (M.20a)

$$a_1 = -\sum \text{finite zeros}$$
 (M.20b)

Also, from Eq. (M.1),

$$n =$$
 number of finite poles – number of finite zeros  
= #finite poles – #finite zeros (M.21)

By examining Eq. (M.16), we conclude that the root locus approaches a straight line as the locus approaches infinity. Further, this straight line intersects the  $\sigma$  axis at

$$\sigma_0 = \frac{\sum \text{ finite poles} - \sum \text{ finite zeros}}{\#\text{finite poles} - \#\text{finite zeros}}$$
(M.22)

which is obtained by substituting Eqs. (M.20)

Let us summarize the results: The root locus approaches straight lines as asymptotes as the locus approaches infinity. Further, the equation of the asymptotes is given by the real-axis intercept and the angle with respect to the real axis as follows:

$$\sigma_0 = \frac{\sum \text{ finite poles} - \sum \text{ finite zeros}}{\#\text{finite poles} - \#\text{finite zeros}}$$
(M.23)

$$\theta = \frac{(2k+1)\pi}{\#\text{finite poles} - \#\text{finite zeros}}$$
(M.24)

where  $k = 0, \pm 1, \pm 2, \pm 3, ...$  Notice that the running index, k, in Eq. (M.24) yields a multiplicity of lines that account for the many branches of a root locus that approach infinity.

## M.2 Derivation of Transition Method for Breakaway and Break-in Points

The *transition* method for finding real-breakaway and break-in points without differentiating can be derived by showing that the natural log of  $1/[G(\sigma)H(\sigma)]$  has a zero derivative at the same value of  $\sigma$  as  $1/[G(\sigma)H(\sigma)]$  (*Franklin, 1991*).

<sup>&</sup>lt;sup>2</sup> Given an *n*th-order polynomial of the form  $s^n + a_{n-1}s^{n-1} + \cdots$ , the coefficient,  $a_{n-1}$ , is the negative sum of the roots of the polynomial.

4

Appendix M: Root Locus Rules: Derivations

We now show that if we work with the natural log we can eliminate the step of differentiation.

First find the derivative of the natural log of  $1/[G(\sigma)H(\sigma)]$  and set it equal to zero. Thus,

$$\frac{d}{d\sigma} \ln\left[\frac{1}{G(\sigma)H(\sigma)}\right] = G(\sigma)H(\sigma)\frac{d}{d\sigma}\left[\frac{1}{G(\sigma)H(\sigma)}\right] = 0$$
(M.25)

Since  $G(\sigma)H(\sigma)$  is not zero at the breakaway or break-in points, letting

$$\frac{d}{d\sigma} \ln \left[ \frac{1}{G(\sigma)H(\sigma)} \right] = 0 \tag{M.26}$$

will thus yield the same value of  $\sigma$  as letting

$$\frac{d}{d\sigma} \left[ \frac{1}{G(\sigma)H(\sigma)} \right] = 0 \tag{M.27}$$

Hence,

$$\frac{d}{d\sigma} \ln\left[\frac{1}{G(\sigma)H(\sigma)}\right] = \frac{d}{d\sigma} \ln\left[\frac{(\sigma+p_1)(\sigma+p_2)\cdots(\sigma+p_n)}{(\sigma+z_1)(\sigma+z_2)\cdots(\sigma+z_m)}\right]$$
$$= \frac{d}{d\sigma} \left[\ln(\sigma+p_1) + \ln(\sigma+p_2)\cdots\ln(\sigma+p_n)\right]$$
$$-\ln(\sigma+z_1) - \ln(\sigma+z_2)\cdots - \ln(\sigma+z_m)\right]$$
$$= \frac{1}{\sigma+p_1} + \frac{1}{\sigma+p_2}\cdots + \frac{1}{\sigma+p_n} - \frac{1}{\sigma+z_1} - \frac{1}{\sigma+z_2}\cdots$$
$$-\frac{1}{\sigma+z_m} = 0$$
(M.28)

or

$$\sum_{i=1}^{n} \frac{1}{\sigma + p_i} = \sum_{i=1}^{m} \frac{1}{\sigma + z_i}$$
(M.29)

where  $z_i$  and  $p_i$  are the negatives of the zero and pole values of G(s)H(s), respectively. Equation (M.29) can be solved for  $\sigma$ , the real axis values that minimize or maximize K, yielding the breakaway and break-in points without differentiating.



Franklin, G. F.; Powell, J. D.; and Emami-Naeini, A. *Feedback Control of Dynamic Systems*, 2d ed. Addison-Wesley, Reading, Mass., 1991.

Kuo, B. Automatic Control Systems, 5th ed. Prentice-Hall, Englewood Cliffs, N.J., 1987.

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